# Applications of Haar basis method for solving some ill-posed inverse problems 

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#### Abstract

In this paper a numerical method consists of combining Haar basis method and Tikhonov regularization method for solving some ill-posed inverse problems using noisy data is presented. By using a sensor located at a point inside the body and measuring the $u(x, t)$ at a point $x=a, 0<a<1$, and applying Haar basis method to the inverse problem, we determine a stable numerical solution to this problem. Results show that an excellent estimation on the unknown functions of the inverse problem can be obtained within a couple of minutes CPU time at pentium IV-2.4 GHz PC.


Keywords Ill-posed inverse problems • Haar basis method •
Tikhonov regularization method • Noisy data
Mathematics Subject Classification (2000) 65M32 • 35K05

## 1 Introduction

Inverse problems appear in many important scientific and technological fields. Hence analysis, design implementation and testing of inverse algorithms are also are great scientific and technological interest.

[^0]Mathematically, the inverse problems belong to the class of problems called the illposed problems, i.e. small errors in the measured data can lead to large deviations in the estimated quantities. As a consequence, their solution does not satisfy the general requirement of existence, uniqueness, and stability under small changes to the input data. To overcome such difficulties, a variety of techniques for solving inverse problems have been proposed $[1-6,14,15,19,20,22-26,30]$ and among the most versatile methods the following can be mentioned: Tikhonov regularization [28], iterative regularization [2], mollification [20], BFM (Base Function Method) [23], SFDM (Semi Finite Difference Method) [19], and the FSM (Function Specification Method) [3].

Beck and Murio [5] presented a new method that combines the function specification method of Beck with the regularization technique of Tikhonov. Murio and Paloschi [21] propose a combined procedure based on a data filtering interpretation of the mollification method and FSM. Beck et al. [3] compare the FSM, the Tikhonov regularization and the iterative regularization, using experimental data.

Zhou et al. [30] investigated the inverse heat conduction problem in a onedimensional composite slab with rate-dependent pyrolysis chemical reaction and outgassing flow effects using the iterative regularization approach. They considered the thermal properties of the temperature-dependent composites.

Huanga et al. [15] applied an iterative regularization method based inverse algorithm in the present study in simultaneously determining the unknown temperature and concentration-dependent heat and mass production rates for a chemically reacting fluid by using interior measurements of temperature and concentration.

Haar functions, [12], have been used from 1910 when they were introduced by the Hungarian mathematician Haar [11]. The Haar transform is one of the earliest of what is known now as a compact, dyadic, orthonormal wavelet transform. The Haar function, being an odd rectangular pulse pair, is the simplest and oldest orthonormal wavelet with compact support. In the mean time, several definitions of the Haar functions and various generalizations have been published and used. They were intended to adopt this concept to some practical applications as well as to extend its in applications to different classes of signals. Haar functions appear very attractive in many applications as for example, image coding, edge extraction and binary logic design.

Recently, Haar wavelets, [12], have been applied extensively for signal processing in communications and physics research, and have proved to be a wonderful mathematical tool. After discretizing the differential equations in a conventional way like the finite difference approximation, wavelets can be used for algebraic manipulations in the system of equations obtained which lead to better condition number of the resulting system.

The previous work, [12], in the system analysis via Haar wavelets was led by Chen and Hsiao [7], who first derived a Haar operational matrix for the integrals of the Haar functions vector and put the application for the Haar analysis into the dynamical systems. Then, the pioneer work in state analysis of linear time delayed systems via Haar wavelets was laid down by Hsiao [13], who first proposed a Haar product matrix and a coefficient matrix. Hsiao and Wang proposed a key idea to transform the timevarying function and its product with states into a Haar product matrix. Kalpana and Balachandar [16] presented Haar basis method of analysis for observer design in the generalized state space or singular system of transistor circuits.

The plan of this paper is as follows. In Sect. 2, we formulate and solve an inverse problem for the heat equation. In addition, a solution of an inverse problem for the wave equation will be discussed. To regularize the resultant ill-conditioned linear system of equations, we apply the Tikhonov regularization (of 0th, 1st and 2nd orders) method to obtain the stable numerical approximation of our solution. Finally some numerical results will be given in Sect. 3 .

## 2 Main results

Definition 2.1 The Haar wavelet family for $x \in[0,1)$ is defined as follows, [12],

$$
h_{i}(x)= \begin{cases}1, & x \in\left[\frac{k}{m}, \frac{k+0.5}{m}\right) \\ -1, & x \in\left[\frac{k+0.5}{m}, \frac{k+1}{m}\right), \\ 0, & \text { elsewhere }\end{cases}
$$

Integer $m=2^{j},(j=0,1, \ldots, J)$ indicates the level of the wavelet; $k=$ $0,1, \ldots, m-1$ is the translation parameter. Maximal level of resolution is $J$. The index $i$ is calculated by $i=m+k+1$; in the case of minimal values $m=1, k=0$ we have $i=2$, the maximal value of $i$ is $i=2^{J+1}=M$. It is assumed that the value $i=1$ corresponds to the scaling function for which $h_{1} \equiv 1$ in $[0,1)$. Let us define the collocation point $x_{l}=\frac{l-0.5}{M},(l=1,2, \ldots, M)$ and discretize the Haar functions $h_{i}(x)$. In this way we get the coefficient matrix $H$ and the operational matrices of integration $P$ and $Q$, which are $M$-square matrices, are defined by the equations

$$
\begin{align*}
(H)_{i l} & =\left(h_{i}\left(x_{l}\right)\right),  \tag{2.1}\\
(P H)_{i l} & =\int_{0}^{x_{l}} h_{i}(x) d x,  \tag{2.2}\\
(Q H)_{i l} & =\int_{0}^{x_{l}} \int_{0}^{x} h_{i}(s) d s d x . \tag{2.3}
\end{align*}
$$

The elements of the matrices $H, P$ and $Q$ can be evaluated by (2.1), (2.2) and (2.3). For example when $M=2$, 4 we have,

$$
\begin{aligned}
& H_{2}=\left(\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right), P_{2}=\frac{1}{4}\left(\begin{array}{cc}
2 & -1 \\
1 & 0
\end{array}\right), Q_{2}=\frac{1}{32}\left(\begin{array}{ll}
5 & -4 \\
4 & -3
\end{array}\right), \\
& H_{4}=\left(\begin{array}{cccc}
1 & 1 & 1 & 1 \\
1 & 1 & -1 & -1 \\
1 & -1 & 0 & 0 \\
0 & 0 & 1 & -1
\end{array}\right), \quad P_{4}=\frac{1}{16}\left(\begin{array}{cccc}
8 & -4 & -2 & -2 \\
4 & 0 & -2 & 2 \\
1 & 1 & 0 & 0 \\
1 & -1 & 0 & 0
\end{array}\right),
\end{aligned}
$$

$$
Q_{4}=\frac{1}{128}\left(\begin{array}{cccc}
21 & -16 & -4 & -12 \\
16 & -11 & -4 & -4 \\
6 & -2 & -3 & 0 \\
2 & -2 & 0 & -3
\end{array}\right)
$$

Remark 1 Any function $f \in L^{2}([0,1))$ can be decomposed as, [12],

$$
\begin{equation*}
f(x)=\sum_{n=1}^{\infty} c_{n} h_{n}(x) \tag{2.4}
\end{equation*}
$$

where the coefficients $c_{n}$ are determined by

$$
c_{n}=2^{j} \int_{0}^{1} f(x) h_{n}(x) d x
$$

where $n=2^{j}+k, j \geqslant 0,0 \leqslant k<2^{j}$.
We should note by Remark 1 that if $f(x)$ is piecewise constant by itself, or may be approximated as piecewise constant during each subinterval, then $f(x)$ will be terminated at finite terms, that is,

$$
\begin{equation*}
f(x)=\sum_{n=1}^{M} c_{n} h_{n}(x)=C_{M}^{T} H_{M}(x), \tag{2.5}
\end{equation*}
$$

where the coefficients $C_{M}^{T}$ and the Haar function vector $H_{M}(x)$ are defined as,

$$
\begin{aligned}
H_{M}(x) & =\left(h_{1}(x) h_{2}(x) \ldots h_{M}(x)\right)^{T}, \\
C_{M}^{T} & =\left(c_{1} c_{2} \ldots c_{M}\right),
\end{aligned}
$$

where ' $T$ ' means transpose and $M=2^{J+1}$.

### 2.1 Inverse problem for the heat equation

One example of the inverse heat conduction problem is the estimation of the heating history experienced by a shuttle or missile reentering the earth's atmosphere from space. The heat flux at the heated surface is needed [3]. To estimate the surface heat flux history, it is necessary to have a mathematical model of the heat transfer process. For example, it is assumed that the section of the skin is of a single material, homogeneous and isotropic, and that it closely approximates a flat plate. Then a possible mathematical model for the temperature in the plate is a one dimensional inverse heat conduction problem as follows, [3]:

$$
\begin{align*}
u_{t}(x, t) & =u_{x x}(x, t), & & 0<x<1, \quad 0<t<t_{f}  \tag{2.6a}\\
u(x, 0) & =\phi(x), & & 0 \leq x \leq 1  \tag{2.6b}\\
u(0, t) & =p(t), & & 0 \leq t \leq t_{f} \\
u(1, t) & =q(t), & & 0 \leq t \leq t_{f} \tag{2.6c}
\end{align*}
$$

and the overspecified condition

$$
\begin{equation*}
u(a, t)=g(t), \quad 0 \leq t \leq t_{f}, \tag{2.6e}
\end{equation*}
$$

where $0<a<1$ is a fixed point, $\phi(x)$ is a continuous known function, $g(t)$ and $q(t)$ are infinitely differentiable known functions and $t_{f}$ represents the final time, while the function $p(t)$ is unknown which remains to be determined from some interior temperature measurements.

Now, let us divide the interval $\left[0, t_{f}\right]$ into $N$ equal parts of length $\Delta t=\frac{t_{f}}{N}$ and denote $t_{s}=(s-1) \Delta t, s=1,2, \ldots, N$. We assume that $\dot{u}^{\prime \prime}$ can be expanded in terms of Haar wavelets as, [12]

$$
\begin{equation*}
\dot{u}^{\prime \prime}(x, t)=\sum_{n=1}^{M} c_{s}(n) h_{n}(x)=C_{M}^{T} H_{M}(x), \tag{2.7}
\end{equation*}
$$

where • and ' mean differentiation with respect to $t$ and $x$, respectively, the vector $C_{M}^{T}$ is constant in each subinterval $\left[t_{s}, t_{s+1}\right], s=1,2, \ldots, N$.

Integrating formula (2.7) with respect to $t$ from $t_{s}$ to $t$ and then twice with respect to $x$ from $a$ to $x$, we obtain

$$
\begin{align*}
u^{\prime \prime}(x, t)= & \left(t-t_{s}\right) C_{M}^{T} H_{M}(x)+u^{\prime \prime}\left(x, t_{s}\right),  \tag{2.8}\\
u(x, t)= & \left(t-t_{s}\right) C_{M}^{T}\left[Q_{M} H_{M}(x)-Q_{M} H_{M}(a)-(x-a) P_{M} H_{M}(a)\right] \\
& \left.+u(a, t)-u\left(a, t_{s}\right)+(x-a)\left[u^{\prime}(a, t)-u^{\prime}\left(a, t_{s}\right)\right)\right]+u\left(x, t_{s}\right),  \tag{2.9}\\
\dot{u}(x, t)= & C_{M}^{T}\left[Q_{M} H_{M}(x)-Q_{M} H_{M}(a)-(x-a) P_{M} H_{M}(a)\right] \\
& +\dot{u}(a, t)+(x-a) \dot{u}^{\prime}(a, t) . \tag{2.10}
\end{align*}
$$

By using the boundary conditions, we obtain

$$
u\left(a, t_{s}\right)=g\left(t_{s}\right), \quad u\left(1, t_{s}\right)=q\left(t_{s}\right), \quad \dot{u}(a, t)=g^{\prime}(t), \quad \dot{u}(1, t)=q^{\prime}(t) .
$$

Putting $x=1$ in (2.9) and (2.10), we obtain

$$
\begin{align*}
u^{\prime}(a, t)-u^{\prime}\left(a, t_{s}\right)= & \frac{t-t_{s}}{a-1} C_{M}^{T}\left[Q_{M} H_{M}(1)-Q_{M} H_{M}(a)-(1-a) P_{M} H_{M}(a)\right] \\
& +\frac{1}{1-a}\left(u(1, t)-u\left(1, t_{s}\right)\right)+\frac{1}{1-a}\left(u\left(a, t_{s}\right)-u(a, t)\right), \tag{2.11}
\end{align*}
$$

$$
\begin{align*}
\dot{u}^{\prime}(a, t)= & \frac{1}{1-a}[\dot{u}(1, t)-\dot{u}(a, t)]-C_{M}^{T}\left[\frac{1}{1-a} Q_{M} H_{M}(1)\right. \\
& \left.-\frac{1}{1-a} Q_{M} H_{M}(a)-P_{M} H_{M}(a)\right] . \tag{2.12}
\end{align*}
$$

Substituting Eqs. (2.11) and (2.12) into Eqs. (2.9) and (2.10), and discretizing the results by assuming $x \rightarrow x_{l}, t \rightarrow t_{s+1}$, we obtain

$$
\begin{align*}
u^{\prime \prime}\left(x_{l}, t_{s+1}\right)= & \left(t_{s+1}-t_{s}\right) C_{M}^{T} H_{M}\left(x_{l}\right)+u^{\prime \prime}\left(x_{l}, t_{s}\right),  \tag{2.13}\\
u\left(x_{l}, t_{s+1}\right)= & \left(t_{s+1}-t_{s}\right) C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)-\frac{x_{l}-a}{1-a} P_{M} F+\frac{x_{l}-1}{1-a} Q_{M} H_{M}(a)\right] \\
& +u\left(x_{l}, t_{s}\right)+\frac{x_{l}-1}{1-a}\left[g\left(t_{s}\right)-g\left(t_{s+1}\right)\right]+\frac{x_{l}-a}{1-a}\left[q\left(t_{s+1}\right)-q\left(t_{s}\right)\right], \\
\dot{u}\left(x_{l}, t_{s+1}\right)= & C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)-\frac{x_{l}-a}{1-a} P_{M} F+\frac{x_{l}-1}{1-a} Q_{M} H_{M}(a)\right]  \tag{2.14}\\
& +\frac{1-x_{l}}{1-a}\left[g^{\prime}\left(t_{s+1}\right)\right]+\frac{x_{l}-a}{1-a}\left[q^{\prime}\left(t_{s+1}\right)\right] . \tag{2.15}
\end{align*}
$$

where the vector $F$ is defined as

$$
F=[1, \underbrace{0, \ldots, 0}_{(M-1)}]^{T}
$$

and $H, P, Q$ are obtained from (2.1)-(2.3).
In the following scheme

$$
\begin{equation*}
\dot{u}\left(x_{l}, t_{s+1}\right)=u^{\prime \prime}\left(x_{l}, t_{s+1}\right), \tag{2.16}
\end{equation*}
$$

which leads us from the time layer $t_{s}$ to $t_{s+1}$ is used where $x_{l}$ is collocation point. Substituting (2.13) and (2.15) into (2.16), we obtain

$$
\begin{align*}
& C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)-\frac{x_{l}-a}{1-a} P_{M} F+\frac{x_{l}-1}{1-a} Q_{M} H_{M}(a)-\Delta t H_{M}\left(x_{l}\right)\right] \\
& \quad=u^{\prime \prime}\left(x_{l}, t_{s}\right)-\frac{1-x_{l}}{1-a} g^{\prime}\left(t_{s+1}\right)-\frac{x_{l}-a}{1-a} q^{\prime}\left(t_{s+1}\right) \tag{2.17}
\end{align*}
$$

Thus the linear system corresponding to the wavelet coefficient $C_{M}^{T}$ can be expressed as

$$
\begin{equation*}
\Lambda \Theta=B \tag{2.18}
\end{equation*}
$$

The Matrix $\Lambda$ is ill-conditioned. On the other hand, as $g(t)$ is affected by measurement errors, the estimate of $\Theta$ by (2.18) will be unstable so that the Tikhonov
regularization method must be used to control this measurement errors. The Tikhonov regularized solution ([10, 17,27] and [28]) to the system of linear algebraic equation (2.18) is given by

$$
\digamma_{\alpha}(\Theta)=\|\Lambda \Theta-B\|_{2}^{2}+\alpha\left\|R^{(s)} \Theta\right\|_{2}^{2}
$$

On the case of the zeroth-, first-, and second-order Tikhonov regularization method the matrix $R^{(s)}$, for $s=0,1,2$, is given by, see e.g. [18]:

$$
\begin{aligned}
& R^{(0)}=I_{M_{1} \times M_{1}} \in \mathbb{R}^{M_{1} \times M_{1}}, \\
& R^{(1)}=\left(\begin{array}{cccccc}
-1 & 1 & \ldots & 0 & 0 & 0 \\
0 & -1 & 1 & \ldots & 0 & 0 \\
\vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\
0 & 0 & \ldots & -1 & 1 & 0 \\
0 & 0 & \ldots & 0 & -1 & 1
\end{array}\right) \in \mathbb{R}^{\left(M_{1}-1\right) \times M_{1}}, \\
& R^{(2)}=\left(\begin{array}{ccccccc}
1 & -2 & 1 & 0 & \ldots & 0 & 0 \\
0 & 1 & -2 & 1 & 0 & \ldots & 0 \\
\vdots & \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \\
0 & 0 & \ldots & 1 & -2 & 1 & 0 \\
0 & 0 & \ldots & 0 & 1 & -2 & 1
\end{array}\right) \in \mathbb{R}^{\left(M_{1}-2\right) \times M_{1}},
\end{aligned}
$$

where $M_{1}=(\gamma+1) \times(\iota+1)$.
Therefore, we obtain the Tikhonov regularized solution of the regularized equation as

$$
\Theta_{\alpha}=\left[\Lambda^{T} \Lambda+\alpha\left(R^{(s)}\right)^{T} R^{(s)}\right]^{-1} \Lambda^{T} B
$$

In our computation, we use the GCV scheme to determine a suitable value of $\alpha$ ([8,9] and [29]).

### 2.2 Inverse problem for the wave equation

In this section, we consider the following inverse problem for the wave equation in the dimensionless form

$$
\begin{align*}
u_{t t}(x, t) & =u_{x x}(x, t), & & 0<x<1, \quad 0<t<t_{f},  \tag{2.19a}\\
u(x, 0) & =f_{1}(x), & & 0 \leq x \leq 1,  \tag{2.19b}\\
u_{t}(x, 0) & =f_{2}(x), & & 0 \leq x \leq 1,  \tag{2.19c}\\
u(0, t) & =p(t), & & 0 \leq t \leq t_{f},  \tag{2.19d}\\
u(1, t) & =q(t), & & 0 \leq t \leq t_{f}, \tag{2.19e}
\end{align*}
$$

and the overspecified condition

$$
\begin{equation*}
u(a, t)=g(t), \quad 0 \leq t \leq t_{f}, \tag{2.19f}
\end{equation*}
$$

where $0<a<1$ is a fixed point, $f_{1}(x)$ is a continuous known function, $g(t)$ and $q(t)$ are infinitely differentiable known functions and $t_{f}$ represents the final time; while the function $p(t)$ is unknown which remains to be determined from some interior temperature measurements.

Now, let us divide the interval $\left[0, t_{f}\right]$ into $N$ equal parts of length $\Delta t=\frac{t_{f}}{N}$ and denote $t_{s}=(s-1) \Delta t, s=1,2, \ldots, N$. We assume by Remark 1 that $\ddot{u}^{\prime \prime}$ can be expanded in terms of $h_{n}$-functions as,

$$
\begin{equation*}
\ddot{u}^{\prime \prime}(x, t)=\sum_{n=1}^{M} c_{s}(n) h_{n}(x)=C_{M}^{T} H_{M}(x), \tag{2.20}
\end{equation*}
$$

where $\cdot=\partial / \partial t$ and $^{\prime}=\partial / \partial x$ and the vector $C_{M}^{T}$ is constant in each subinterval $\left[t_{s}, t_{s+1}\right], s=1,2, \ldots, N$.

Integrating formula (2.20) twice with respect to $t$ from $t_{s}$ to $t$ and then twice with respect to $x$ from $a$ to $x$, we obtain

$$
\begin{align*}
u^{\prime \prime}(x, t)= & \frac{1}{2}\left(t^{2}+t_{s}^{2}-2 t t_{s}\right) C_{M}^{T} H_{M}(x)+u^{\prime \prime}\left(x, t_{s}\right)+\left(t-t_{s}\right) \dot{u}^{\prime \prime}\left(x, t_{s}\right),  \tag{2.21}\\
\ddot{u}(x, t)= & C_{M}^{T}\left[Q_{M} H_{M}(x)-Q_{M} H_{M}(a)-(x-a) P_{M} H_{M}(a)\right] \\
& +\ddot{u}(a, t)+(x-a) \ddot{u}^{\prime}(a, t),  \tag{2.22}\\
u(x, t)= & \frac{1}{2}\left(t^{2}+t_{s}^{2}-2 t t_{s}\right) C_{M}^{T}\left[Q_{M} H_{M}(x)-Q_{M} H_{M}(a)-(x-a) P_{M} H_{M}(a)\right] \\
& +u\left(x, t_{s}\right)+\left(t-t_{s}\right) \dot{u}\left(x, t_{s}\right)+u(a, t)-u\left(a, t_{s}\right)-\left(t-t_{s}\right) \dot{u}\left(a, t_{s}\right) \\
& +(x-a)\left[u^{\prime}(a, t)-u^{\prime}\left(a, t_{s}\right)-\left(t-t_{s}\right) \dot{u}^{\prime}\left(a, t_{s}\right)\right] . \tag{2.23}
\end{align*}
$$

By using the boundary conditions, we obtain

$$
\begin{aligned}
u\left(1, t_{s}\right) & =q\left(t_{s}\right), \quad u\left(a, t_{s}\right)=g\left(t_{s}\right), \\
\dot{u}\left(1, t_{s}\right) & =q^{\prime}\left(t_{s}\right), \quad \dot{u}\left(a, t_{s}\right)=g^{\prime}\left(t_{s}\right), \\
\ddot{u}(1, t) & =q^{\prime \prime}(t), \quad \ddot{u}(a, t)=g^{\prime \prime}(t) .
\end{aligned}
$$

Putting $x=1$ in (2.22) and (2.23), we obtain

$$
\begin{align*}
& \ddot{u}^{\prime}(a, t)=C_{M}^{T}\left[\frac{1}{a-1} Q_{M} H_{M}(1)\right. \\
& \left.-\frac{1}{a-1} Q_{M} H_{M}(a)+P_{M} H_{M}(a)\right]+\frac{1}{1-a}\left[q^{\prime \prime}(t)-g^{\prime \prime}(t)\right],  \tag{2.24}\\
& u^{\prime}(a, t)-u^{\prime}\left(a, t_{s}\right)-\left(t-t_{s}\right) \dot{u}^{\prime}\left(a, t_{s}\right)=\frac{1}{1-a}\left[q(t)-q\left(t_{s}\right)\right] \\
& +\frac{1}{1-a}\left[g\left(t_{s}\right)-g(t)\right]+\frac{t-t_{s}}{1-a}\left[g^{\prime}\left(t_{s}\right)-q^{\prime}\left(t_{s}\right)\right] \\
& +\frac{1}{2}\left(t^{2}+t_{s}^{2}-2 t t_{s}\right) C_{M}^{T}\left[\frac{1}{a-1} Q_{M} H_{M}(1)-\frac{1}{a-1} Q_{M} H_{M}(a)+P_{M} H_{M}(a)\right] . \tag{2.25}
\end{align*}
$$

Substituting (2.24) and (2.25) into (2.22) and (2.23), and discretizing the results by assuming $x \rightarrow x_{l}, t \rightarrow t_{s+1}$ we obtain

$$
\begin{align*}
u^{\prime \prime}\left(x_{l}, t_{s+1}\right)= & \frac{1}{2}\left(t_{s+1}^{2}+t_{s}^{2}-2 t_{s+1} t_{s}\right) C_{M}^{T} H_{M}\left(x_{l}\right) \\
& +u^{\prime \prime}\left(x_{l}, t_{s}\right)+\left(t_{s+1}-t_{s}\right) \dot{u}^{\prime \prime}\left(x_{l}, t_{s}\right),  \tag{2.26}\\
\dot{u}^{\prime \prime}\left(x_{l}, t_{s+1}\right)= & \left(t_{s+1}-t_{s}\right) C_{M}^{T} H_{M}\left(x_{l}\right)+\dot{u}^{\prime \prime}\left(x_{l}, t_{s}\right),  \tag{2.27}\\
\ddot{u}\left(x_{l}, t_{s+1}\right)= & C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)+\frac{1-x_{l}}{a-1} Q_{M} H_{M}(a)+\frac{x_{l}-a}{a-1} P_{M} F\right] \\
& +\frac{1-x_{l}}{1-a} g^{\prime \prime}\left(t_{s+1}\right)+\frac{x_{l}-a}{1-a} q^{\prime \prime}\left(t_{s+1}\right),  \tag{2.28}\\
u\left(x_{l}, t_{s+1}\right)= & \frac{1}{2}\left(t_{s+1}^{2}+t_{s}^{2}-2 t_{s+1} t_{s}\right) C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)+\frac{1-x_{l}}{a-1} Q_{M} H_{M}(a)\right. \\
& \left.+\frac{x_{l}-a}{a-1} P_{M} F\right]+u\left(x_{l}, t_{s}\right)+\left(t_{s+1}-t_{s}\right) \dot{u}\left(x_{l}, t_{s}\right) \\
& +\frac{x_{l}-1}{1-a}\left[g\left(t_{s}\right)-g\left(t_{s+1}\right)\right]+\frac{x_{l}-a}{1-a}\left[q\left(t_{s+1}\right)-q\left(t_{s}\right)\right] \\
& +\frac{t_{s+1}-t_{s}}{1-a}\left[\left(x_{l}-1\right) g^{\prime}\left(t_{s}\right)-\left(x_{l}-a\right) q^{\prime}\left(t_{s}\right)\right],  \tag{2.29}\\
\dot{u}\left(x_{l}, t_{s+1}\right)= & \left(t_{s+1}-t_{s}\right) C_{M}^{T}\left[Q_{M} H_{M}\left(x_{l}\right)+\frac{1-x_{l}}{a-1} Q_{M} H_{M}(a)+\frac{x_{l}-a}{a-1} P_{M} F\right] \\
& +\dot{u}\left(x_{l}, t_{s}\right)+\frac{1-x_{l}}{1-a}\left[g^{\prime}\left(t_{s+1}\right)-g^{\prime}\left(t_{s}\right)\right]+\frac{x_{l}-a}{1-a}\left[q^{\prime}\left(t_{s+1}\right)-q^{\prime}\left(t_{s}\right)\right] . \tag{2.30}
\end{align*}
$$

where the vector $F$ is defined as

$$
F=[1, \underbrace{0, \ldots, 0}_{(M-1)}]^{T}
$$

and $H, P$ and $Q$ are obtained by (2.1), (2.2) and (2.3). In the following scheme

$$
\begin{equation*}
\ddot{u}\left(x_{l}, t_{s+1}\right)=u^{\prime \prime}\left(x_{l}, t_{s+1}\right), \tag{2.31}
\end{equation*}
$$

which leads us from the time layer $t_{s}$ to $t_{s+1}$ is used where $x_{l}$ is collocation point. Substituting Eqs. (2.26) and (2.28) into Eq. (2.31), we obtain

$$
\begin{align*}
C_{M}^{T} & {\left[Q_{M} H_{M}\left(x_{l}\right)+\frac{1-x_{l}}{a-1} Q_{M} H_{M}(a)+\frac{x_{l}-a}{a-1} P_{M} F\right.} \\
& \left.-\frac{1}{2}\left(t_{s+1}^{2}+t_{s}^{2}-2 t_{s+1} t_{s}\right) H_{M}\left(x_{l}\right)\right]=u^{\prime \prime}\left(x_{l}, t_{s}\right)+\Delta t \dot{u}^{\prime \prime}\left(x_{l}, t_{s}\right) \\
& -\frac{1-x_{l}}{1-a} g^{\prime \prime}\left(t_{s+1}\right)-\frac{x_{l}-a}{1-a} q^{\prime \prime}\left(t_{s+1}\right) . \tag{2.32}
\end{align*}
$$

From the formula (2.32) the wavelet coefficient $C_{M}^{T}$ can be calculated.
In matrix form, the wavelet coefficient $C_{M}^{T}$, can be obtained from solving the following matrix equation

$$
\begin{equation*}
A \lambda=b . \tag{2.33}
\end{equation*}
$$

Similarly, the Tikhonov regularized solution to the system of linear algebraic equation (2.18) is given by (see e.g. [10,17] and [27])

$$
\lambda_{\alpha}=\left[A^{T} A+\alpha\left(R^{(s)}\right)^{T} R^{(s)}\right]^{-1} A^{T} b .
$$

Table 1 The comparison between exact and Tikhonov solutions of $p(t)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0.01 | 1.921879 | 1.919898 | 1.922970 | 1.922109 |
| 0.02 | 1.847433 | 1.844146 | 1.848860 | 1.847618 |
| 0.1 | 1.370640 | 1.361024 | 1.370366 | 1.369286 |
| 0.11 | 1.324373 | 1.314032 | 1.323874 | 1.322876 |
| 0.5 | 1.020671 | 0.991988 | 1.017550 | 1.017308 |
| 0.51 | 1.040357 | 1.011336 | 1.037165 | 1.036948 |
| 0.8 | 2.001524 | 1.962028 | 1.998064 | 1.997799 |
| 0.81 | 2.046628 | 2.006747 | 2.043149 | 2.042928 |
| 0.9 | 2.484647 | 2.44109 | 2.481159 | 2.480956 |
| 0.91 | 2.536805 | 2.492833 | 2.533312 | 2.533111 |
| 1 | 3.036631 | 2.988876 | 3.033070 | 3.033044 |
|  | $S$ | $3.053 \mathrm{e}-002$ | $2.875 \mathrm{e}-003$ | $3.129 \mathrm{e}-003$ |

Table 2 The comparison between exact and Tikhonov solutions of $u(0.2, t)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0.01 | 2.520089 | 2.521307 | 2.519971 | 2.520238 |
| 0.02 | 2.423450 | 2.425538 | 2.423526 | 2.423825 |
| 0.1 | 1.799281 | 1.806399 | 1.801775 | 1.801910 |
| 0.11 | 1.737893 | 1.745568 | 1.740668 | 1.740750 |
| 0.5 | 1.165108 | 1.186168 | 1.171230 | 1.171247 |
| 0.51 | 1.182700 | 1.203986 | 1.188859 | 1.188836 |
| 0.8 | 2.123236 | 2.151417 | 2.129903 | 2.129798 |
| 0.81 | 2.168547 | 2.196991 | 2.175235 | 2.175103 |
| 0.9 | 2.610014 | 2.640830 | 2.616710 | 2.616683 |
| 0.91 | 2.662706 | 2.693775 | 2.669420 | 2.669383 |
| 1 | 3.168405 | 3.201933 | 3.175197 | 3.175122 |
|  | $S$ | $2.199 \mathrm{e}-002$ | $5.703 \mathrm{e}-003$ | $5.652 \mathrm{e}-003$ |

Table 3 The comparison between exact and Tikhonov solutions of $u(x, 0.5)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 1.020671 | 0.991984 | 1.017502 | 1.017258 |
| 0.1 | 1.084074 | 1.084074 | 1.084074 | 1.084074 |
| 0.2 | 1.165108 | 1.186201 | 1.171323 | 1.171244 |
| 0.3 | 1.263251 | 1.291492 | 1.271317 | 1.270897 |
| 0.4 | 1.379145 | 1.411504 | 1.389255 | 1.388597 |
| 0.5 | 1.514630 | 1.549149 | 1.526990 | 1.526255 |
| 0.6 | 1.672755 | 1.704997 | 1.684498 | 1.683785 |
| 0.7 | 1.857762 | 1.884610 | 1.866750 | 1.866089 |
| 0.8 | 2.075052 | 2.098842 | 2.084764 | 2.084199 |
| 0.9 | 2.331120 | 2.342195 | 2.334015 | 2.333667 |
| 1 | 2.633481 | 2.633481 | 2.633481 | 2.633481 |
|  | $S$ | $2.598 \mathrm{e}-002$ | $8.307 \mathrm{e}-003$ | $7.831 \mathrm{e}-003$ |

## 3 Numerical results and discussion

In this section, we are going to study numerically the inverse problems (2.6) and (2.19) with the unknown boundary condition. The main aim here is to show the applicability of the present method, described in Sect. 2, for solving the inverse problems (2.6) and (2.19). As expected the inverse problems are ill-posed and therefore it is necessary to investigate the stability of the present method by giving a test problem.

Remark 2 In an inverse problem there are two sources of error in the estimation; the first source is the unavoidable bias deviation, and the second source of error is the variance due to the amplification of measurement errors, [6].


Fig. 1 Difference between the $p(t)_{\text {Exact }}$ and $p(t)_{0 \text { th order Tikhonov }}$ of problem (2.6) with noisy data


Fig. 2 Difference between the $p(t)_{\text {Exact }}$ and $p(t)_{1 \text { st order Tikhonov }}$ of problem (2.6) with noisy data

Therefore, we compare exact and approximate solutions by considering total error $S$ defined by

$$
\begin{equation*}
S=\left[\frac{1}{N-1} \sum_{i=1}^{N}\left(\widehat{\Phi_{i}}-\Phi_{i}\right)^{2}\right]^{\frac{1}{2}} \tag{3.1}
\end{equation*}
$$

where $N, \Phi$ and $\widehat{\Phi}$ are the number of estimated values, the estimated values and the exact values, respectively.

Example 3.1 In this example we solve the problem (2.6) with given data,

$$
\begin{aligned}
& u(x, 0)=2(\sin (2 x)+\cos (2 x))+\frac{1}{4} x^{4}, \quad 0 \leq x \leq 1 \\
& u(1, t)=2 e^{-4 t}(\sin 2+\cos 2)+3\left(t^{2}+t+\frac{1}{12}\right), \quad 0 \leq t \leq t_{f}
\end{aligned}
$$



Fig. 3 Difference between the $p(t)$ Exact and $p(t)_{2 \text { nd order Tikhonov of problem (2.6) with noisy data }}$


Fig. 4 Difference between the $u(0.2, t)_{\text {Exact }}$ and $u(0.2, t)_{0 \text { th order Tikhonov }}$ of problem (2.6) with noisy data

$$
u(0.1, t)=2 e^{-4 t}(\sin 0.2+\cos 0.2)+3\left(t^{2}+(0.01) t+\frac{0.0001}{12}\right), \quad 0 \leq t \leq t_{f}
$$

The exact solution of this problem is

$$
u(x, t)=2 e^{-4 t}(\sin (2 x)+\cos (2 x))+3\left(t^{2}+t x^{2}+\frac{1}{12} x^{4}\right)
$$

Our results obtained for $p(t)=u(0, t), u(0.2, t)$ and $u(x, 0.5)$ when $t_{f}=1, \Delta t=$ 0.01 and $\Delta x=\frac{1}{4}$ with noisy data (noisy data $=$ input data $+(0.01) \operatorname{rand}(1)$ ) are presented in Tables 1, 2 and 3 and Figs. 1, 2, 3, 4, 5, 6, 7, 8 and 9.


Fig. 5 Difference between the $u(0.2, t)$ Exact $a n d(0.2, t)$ st order Tikhonov of problem (2.6) with noisy data


Fig. 6 Difference between the $u(0.2, t)_{\text {Exact }}$ and $u(0.2, t)_{2 \text { nd order Tikhonov }}$ of problem (2.6) with noisy data


Fig. 7 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{0 \text { th order Tikhonov }}$ of problem (2.6) with noisy data

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Fig. 8 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{1 \text { st }}$ order Tikhonov of problem (2.6) with noisy data


Fig. 9 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{2 \text { nd order Tikhonov }}$ of problem (2.6) with noisy data

Example 3.2 In this example we solve the problem (2.19) with given data,

$$
\begin{aligned}
u(x, 0) & =e^{-x}+x^{2}, \quad 0 \leq x \leq 1 \\
u_{t}(x, 0) & =-e^{-x}+x^{3}, \quad 0 \leq x \leq 1 \\
u(1, t) & =e^{-1-t}+t+t^{3}+t^{2}+1, \quad 0 \leq t \leq t_{f} \\
u(0.1, t) & =e^{-0.1-t}+(0.001) t+(0.1) t^{3}+t^{2}+0.01, \quad 0 \leq t \leq t_{f}
\end{aligned}
$$

The exact solution of this problem is

$$
u(x, t)=e^{-x-t}+x^{3} t+x t^{3}+t^{2}+x^{2} .
$$

Table 4 The comparison between exact and Tikhonov solutions of $p(t)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0.01 | 0.990150 | 0.990149 | 0.990150 | 0.990150 |
| 0.02 | 0.980599 | 0.980596 | 0.980599 | 0.980598 |
| 0.1 | 0.914837 | 0.914780 | 0.914836 | 0.914825 |
| 0.11 | 0.907934 | 0.907865 | 0.907932 | 0.907918 |
| 0.5 | 0.856531 | 0.855188 | 0.856137 | 0.856094 |
| 0.51 | 0.860596 | 0.859197 | 0.860186 | 0.860145 |
| 0.8 | 1.089329 | 1.085894 | 1.088756 | 1.088546 |
| 0.81 | 1.100958 | 1.097442 | 1.100387 | 1.100166 |
| 0.9 | 1.216570 | 1.212314 | 1.216003 | 1.215702 |
| 0.91 | 1.230624 | 1.226286 | 1.230056 | 1.229749 |
| 1 | 1.367879 | 1.362790 | 1.367285 | 1.366945 |
|  | $S$ | $2.402 \mathrm{e}-003$ | $4.086 \mathrm{e}-004$ | $5.448 \mathrm{e}-004$ |

Table 5 The comparison between exact and Tikhonov solutions of $u(0.2, t)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0.01 | 0.850764 | 0.850765 | 0.850765 | 0.850765 |
| 0.02 | 0.843080 | 0.843082 | 0.843081 | 0.843081 |
| 0.1 | 0.791818 | 0.791850 | 0.791824 | 0.791831 |
| 0.11 | 0.786693 | 0.786731 | 0.786701 | 0.786708 |
| 0.5 | 0.815585 | 0.816291 | 0.815888 | 0.815874 |
| 0.51 | 0.822354 | 0.823091 | 0.822669 | 0.822650 |
| 0.8 | 1.156679 | 1.158589 | 1.157098 | 1.157077 |
| 0.81 | 1.173087 | 1.175043 | 1.173501 | 1.173487 |
| 0.9 | 1.335871 | 1.338252 | 1.336227 | 1.336287 |
| 0.91 | 1.355653 | 1.358082 | 1.356002 | 1.356071 |
| 1 | 1.549194 | 1.552046 | 1.549486 | 1.549625 |
|  | $S$ | $1.332 \mathrm{e}-003$ | $2.907 \mathrm{e}-004$ | $2.910 \mathrm{e}-004$ |

Our results obtained for $u(0, t), u(0.2, t)$ and $u(x, t)$ when $t_{f}=1, \Delta t=0.01$ and $\Delta x=\frac{1}{4}$ with noisy data (noisy data $=$ input data $+(0.01)$ rand(1)) are presented in Tables 4, 5 and 6 and Figs. 10, 11, 12, 13, 14, 15, 16, 17 and 18.

## 4 Conclusion

A numerical method, to estimate unknown boundary condition is proposed for two types of the inverse problems, the heat problem (2.6) and the wave problem (2.19), by using Haar basis method. The following results are obtained.

1. The present study, successfully applies the numerical method to inverse problems.

Table 6 The comparison between exact and Tikhonov solutions of $u(x, 0.5)$ with noisy data

| $t$ | Exact | 0th order Tikhonov | 1st order Tikhonov | 2nd order Tikhonov |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0.856531 | 0.855211 | 0.856114 | 0.856076 |
| 0.1 | 0.821812 | 0.821812 | 0.821812 | 0.821812 |
| 0.2 | 0.815585 | 0.816278 | 0.815838 | 0.815847 |
| 0.3 | 0.840329 | 0.841219 | 0.840752 | 0.840744 |
| 0.4 | 0.898570 | 0.899482 | 0.899091 | 0.899071 |
| 0.5 | 0.992879 | 0.993755 | 0.993490 | 0.993469 |
| 0.6 | 1.125871 | 1.126669 | 1.126523 | 1.126504 |
| 0.7 | 1.300194 | 1.300887 | 1.300791 | 1.300772 |
| 0.8 | 1.518532 | 1.519118 | 1.519017 | 1.518998 |
| 0.9 | 1.783597 | 1.783947 | 1.783873 | 1.783860 |
| 1 | 2.098130 | 2.098130 | 2.098130 | 2.098130 |
|  | $S$ | $7.696 \mathrm{e}-004$ | $4.641 \mathrm{e}-004$ | $4.545 \mathrm{e}-004$ |



Fig. 10 Difference between the $p(t)_{\text {Exact }}$ and $p(t)_{\text {Oth order Tikhonov }}$ of problem (2.19) with noisy data
2. Numerical results show that an excellent estimation can be obtained within a couple of minutes CPU time at pentium(R) 4 CPU 3.20 GHz .
3. The present method has been found stable with respect to small perturbation in the input data.
4. Numerical results show that our approximations of unknown function using the (1st and 2nd order) Tikhonov regularization combined with the Haar basis method, are more accurate than those obtained by the 0th order Tikhonov regularization with noisy data.


Fig. 11 Difference between the $p(t)_{\text {Exact }}$ and $p(t)_{1 \text { st order Tikhonov }}$ of problem (2.19) with noisy data


Fig. 12 Difference between the $p(t)_{\text {Exact }}$ and $p(t)_{2 \text { nd order Tikhonov }}$ of problem (2.19) with noisy data


Fig. 13 Difference between the $u(0.2, t)_{\text {Exact }}$ and $u(0.2, t)_{\text {Oth order Tikhonov }}$ of problem (2.19) with noisy data


Fig. 14 Difference between the $u(0.2, t)_{\text {Exact }}$ and $u(0.2, t)_{1 \text { st order Tikhonov of problem (2.19) with noisy }}$ data


Fig. 15 Difference between the $u(0.2, t)_{\text {Exact }}$ and $u(0.2, t)_{2 \text { nd order Tikhonov of problem (2.19) with noisy }}$ data


Fig. 16 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{\text {Oth order Tikhonov }}$ of problem (2.19) with noisy data


Fig. 17 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{1 \text { st order Tikhonov }}$ of problem (2.19) with noisy data


Fig. 18 Difference between the $u(x, 0.5)_{\text {Exact }}$ and $u(x, 0.5)_{2 \text { nd orderTikhonov }}$ of problem (2.19) with noisy data

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